## 150mA ULTRA LOW DROPOUT, LOW NOISE REGULATOR

March 1, 2000

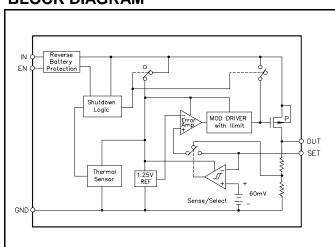
TEL:805-498-2111 FAX:805-498-3804 WEB:http://www.semtech.com

#### **DESCRIPTION**

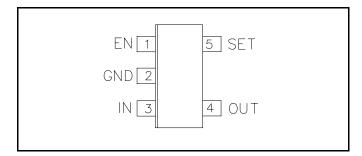
The SC8863 is a low dropout linear regulator that operates from a +2.5V to +5.5V input range and delivers up to 150mA. A PMOS pass transistor allows the low 110 $\mu$ A supply current to remain independent of load, making these devices ideal for battery operated portable equipment such as cellular phones, cordless phones and personal digital assistants.

The SC8863 output voltage can be preset or adjusted with an external resistor divider. Other features include low powered shutdown, short circuit protection, thermal shutdown protection and reverse battery protection. The SC8863 comes in the tiny 5 lead SOT-23 package.

### **BLOCK DIAGRAM**



#### PIN CONFIGURATION



#### **FEATURES**

- Guaranteed 150 mA output current
- Fixed or adjustable output
- Very small external components
- Low 75μV<sub>RMS</sub> output noise
- Very low supply current
- Thermal overload protection
- Reverse battery protection
- Low power shutdown
- Surface mount packaging (5 pin SOT-23)
- Full industrial temperature range

#### **APPLICATIONS**

- Battery Powered Systems
- Cellular Telephones
- Cordless Telephones
- Personal Digital Assistants
- Portable Instrumentation
- Modems
- PCMCIA cards

#### ORDERING INFORMATION

Part Number	Package			
SC8863-XXXCSK <sup>(1)(2)</sup>	SOT-23-5			

#### Notes:

- (1) Where -XXX denotes voltage options. Available voltages are: 2.50V (-250), 2.80V (-280), 3.00V (-300), and 3.30 (-330)V.
- (2) Add suffix 'TR' for tape and reel.

#### **ABSOLUTE MAXIMUM RATINGS**

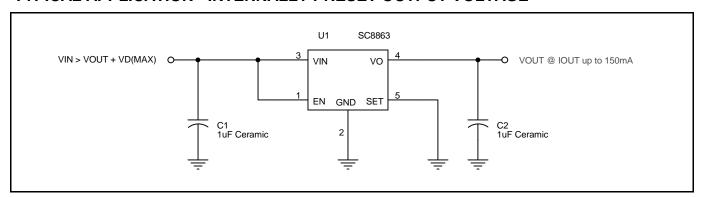
Parameter	Symbol	Maximum	Units	
Input Supply Voltage	V <sub>IN</sub>	-0.6 to +7	V	
Power Dissipation T <sub>A</sub> = 70°C	P <sub>D</sub>	571	mW	
Thermal Resistance	$\theta_{JA}$	256	°C/W	
	$\theta_{JC}$	81	°C/W	
Operating Ambient Temperature Range	T <sub>A</sub>	-40 to +85	°C	
Operating Junction Temperature Range	TJ	-40 to +150	°C	
Storage Temperature Range	T <sub>STG</sub>	-65 to +150	°C	
Lead Temperature (Soldering) 10 Sec	T <sub>LEAD</sub>	+300	°C	
ESD Rating	ESD	1.25	kV	



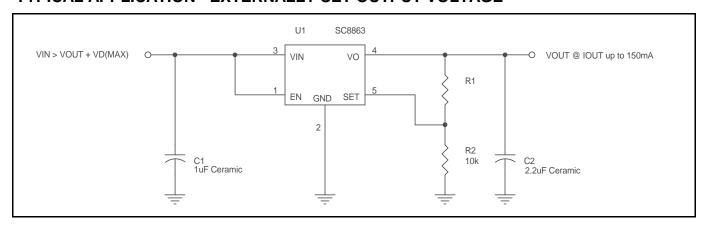
#### **PIN DESCRIPTIONS**

Pin Name	Description			
N	Active high enable pin. Connect to VIN if not being used.			
GND	Ground pin. Can be used for heatsinking if needed.			
N	Input pin.			
DUT	Regulator output, sourcing up to 150mA			
	Connecting this pin to ground results in the internally preset value for $V_{OUT}$ . Connecting to an external resistor divider changes $V_{OUT}$ to: $V_{OUT} = 1.250 \cdot \left(1 + \frac{R1}{R2}\right)$			
113	N ND I UT ET			

#### TYPICAL APPLICATION - INTERNALLY PRESET OUTPUT VOLTAGE



### TYPICAL APPLICATION - EXTERNALLY SET OUTPUT VOLTAGE(1)



#### NOTES:

(1) Select R1 and R2 such that the current flowing through them is  $\geq 10\mu A$  (i.e.  $R2 \leq 120k\Omega$ ). A value of  $10k\Omega$  is recommended for R2. Please see Component Selection - Externally Set Output.



# 150mA ULTRA LOW DROPOUT, LOW NOISE REGULATOR

March 1, 2000

#### **ELECTRICAL CHARACTERISTICS**

Unless specified:  $V_{IN} = 3.6V$ ,  $V_{SET} = GND$ ,  $V_{EN} = V_{IN}$ ,  $T_A = 25^{\circ}C$ . Values in **bold** apply over full operating ambient temperature range.

Parameter	Symbol	Conditions	Min	Тур	Max	Units
IN	1					
Supply Voltage Range	V <sub>IN</sub>		2.5		5.5	V
Supply Current	IQ	$I_{OUT} = OmA$		100	130	μA
					160	
		$50\text{mA} \le I_{\text{OUT}} \le 150\text{mA}$		110	160	μΑ
					200	
		V <sub>EN</sub> = 0V		0.0001	1	μA
					2	
OUT				i.	i.	
Output Voltage (1)	$V_{OUT}$	I <sub>OUT</sub> = 1mA	-2.0%	V <sub>OUT</sub>	+2.0%	V
		$1\text{mA} \le I_{\text{OUT}} \le 150\text{mA}, \ V_{\text{OUT}} + 1V \le V_{\text{IN}} \le 5.5V$	-3.5%		+3.5%	
Line Regulation <sup>(1)</sup>	REG <sub>(LINE)</sub>	$2.5V \le V_{IN} \le 5.5V$ , SET = OUT, $I_{OUT} = 1$ mA		5	10	mV
					12	ı
Load Regulation <sup>(1)</sup>	REG <sub>(LOAD)</sub>	I <sub>OUT</sub> = 0mA to 50mA		-10	-15	mV
					-20	
		I <sub>OUT</sub> = 0mA to 100mA		-15	-20	mV
					-25	
		I <sub>OUT</sub> = 0mA to 50mA, SET = OUT		-2.5	-7.5	mV
					-15.0	
		I <sub>OUT</sub> = 0mA to 100mA, SET = OUT		-5	-15	mV
					-30	
Dropout Voltage(1)(2)	t Voltage <sup>(1)(2)</sup> V <sub>D</sub>	$I_{OUT} = 1mA$		1.1		mV
		$I_{OUT} = 50mA$		55	90	mV
					120	
		$I_{OUT} = 100 \text{mA}$		110	180	mV
					240	
Current Limit	I <sub>LIM</sub>		150	240	350	mA
Output Voltage Noise	e <sub>n</sub>	10Hz to 99kHz, $I_{OUT} = 50$ mA, $C_{OUT} = 1$ $\mu$ F		90		$\mu V_{\text{RMS}}$
		$C_{OUT} = 100 \mu F$		75		
Power Supply Rejection Ratio	PSRR	f ≤ 1kHz		55		dB



### **ELECTRICAL CHARACTERISTICS (Cont.)**

Unless specified:  $V_{IN} = 3.6V$ ,  $V_{SET} = GND$ ,  $V_{EN} = V_{IN}$ ,  $T_A = 25^{\circ}C$ . Values in **bold** apply over full operating ambient temperature range.

Parameter	Symbol	Conditions	Min	Тур	Max	Units
EN						
EN Input Threshold	V <sub>IH</sub>		1.8			V
	V <sub>IL</sub>				0.4	
EN Input Bias Current <sup>(3)</sup>	I <sub>EN</sub>	$EN = V_{IN}$		0	100	nA
					200	
SET						
Sense/Select Threshold	$V_{TH}$		20	55	80	mV
SET Reference Voltage <sup>(1)</sup>	V <sub>SET</sub>	I <sub>OUT</sub> = 1mA	1.225	1.250	1.275	V
		1mA $\leq I_{OUT} \leq 150$ mA, $2.5$ V $\leq V_{IN} \leq 5.5$ V	1.206		1.294	
SET Input Leakage Current <sup>(3)</sup>	I <sub>SET</sub>	SET = 1.3V		0.015	2.500	nA
					5.000	
OVER TEMPERATURE PROT	ECTION		I		1	1
High Trip Level	T <sub>HI</sub>			170		°C
Hysteresis	T <sub>HYST</sub>			10		°C

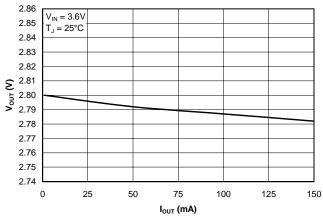
#### NOTE:

- (1) Low duty cycle pulse resting with Kelvin corrections required.
- (2) Defined as the input to output differential at which the output voltage drops 100mV below the value measured at a differential of 2V.
- (3) Guaranteed by design.

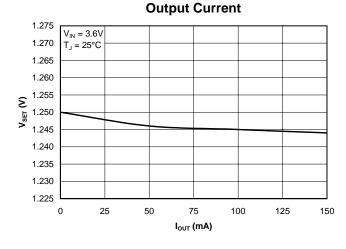


### **TYPICAL CHARACTERISTICS**

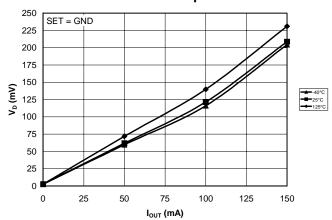
### **Output Voltage vs. Output Current**



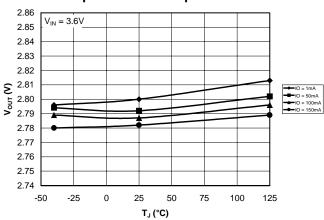
## SET Reference Voltage vs.



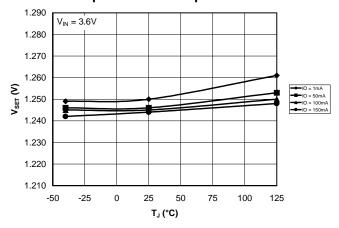
## Dropout Voltage vs. Output Current vs. Junction Temperature



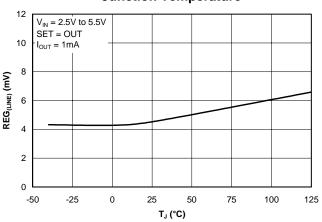
#### Output Voltage vs. Junction Temperature vs. Output Current



#### SET Reference Voltage vs. Junction Temperature vs. Output Current



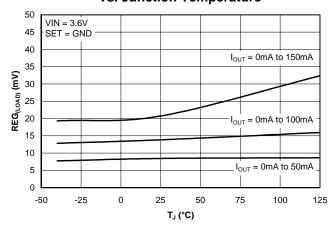
## Line Regulation vs. Junction Temperature



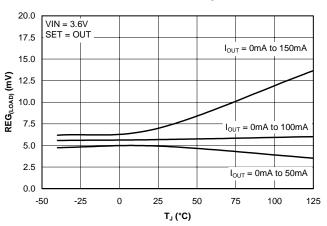


### **TYPICAL CHARACTERISTICS (Cont.)**

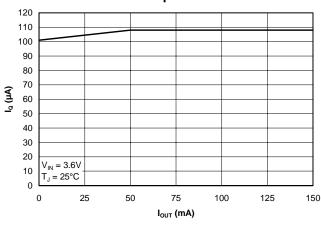
## Load Regulation (SET = GND) vs. Junction Temperature



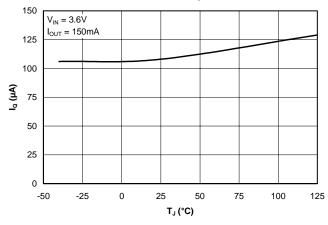
## Load Regulation (SET = OUT) vs. Junction Temperature



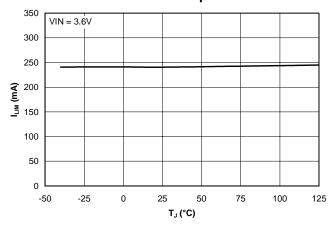
## Supply Current vs. Output Current



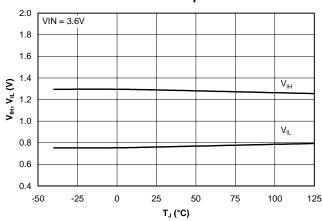
Supply Current vs. Junction Temperature



#### Current Limit vs. Junction Temperature

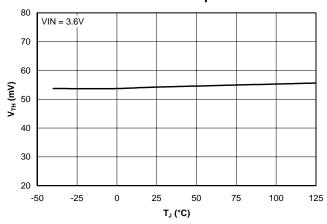


## Enable Input Threshold vs. Junction Temperature



#### **TYPICAL CHARACTERISTICS (Cont.)**

## Sense/Select Threshold vs. Junction Temperature



#### APPLICATIONS INFORMATION

#### **Theory Of Operation**

The SC8863 is intended for applications where very low dropout voltage, low supply current and low output noise are critical. It provides a very simple, low cost solution that uses very little pcb real estate. Fixed output voltage options require the use of only two external capacitors for operation.

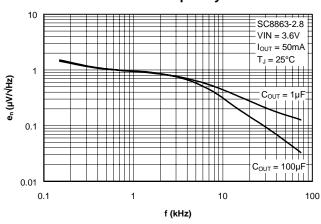
Each voltage option has both fixed and adjustable output voltage modes. Grounding the SET pin (pulling it below the Sense/Select threshold of 55mV) will connect the internal resistor divider to the error amplifier resulting with the internally preset output voltage. If SET is pulled above this threshold, then the Sense/ Select switch will connect the SET pin to the error amplifier. The output will be regulated such that the voltage at SET will equal  $V_{\text{SET}}$ , the SET reference voltage (typically 1.250V).

An active high enable pin (EN) is provided to allow the customer to shut down the part and enter an extremely low power Off-state. A logic Low signal will reduce the supply current to 0.1nA.

#### **Component Selection - General**

Output capacitor - Semtech recommends a minimum capacitance of 1µF at the output with an equivalent series resistance (ESR) of <1 $\Omega$  over temperature. Ceramic capacitors are ideal for this application. Increas-

## Output Spectral Noise Density vs. Frequency



ing the bulk capacitance will further reduce output noise and improve the overall transient response.

Input capacitor - Semtech recommends the use of a  $1\mu F$  ceramic capacitor at the input. This allows for the device being some distance from any bulk capacitance on the rail. Additionally, input droop due to load transients is reduced, improving load transient response.

#### **Component Selection - Externally Set Output**

Please refer to Figure 1 below. The output voltage can be externally adjusted anywhere within the range from 1.25V to  $(VIN_{(MIN)} - VD_{(MAX)})$ . The output voltage will be in accordance with the following equation:

$$VO = 1.250 \bullet \left(1 + \frac{R1}{R2}\right)$$

1% tolerance resistors are recommended. The values of R1 and R2 should be selected such that the current flow through them is  $\geq 10\mu A$  (thus R2  $\leq 120k\Omega)$ . At high input voltages and/or high output currents, sta-

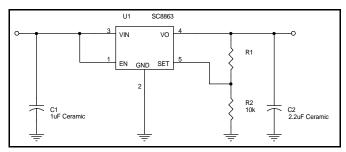


Figure 1: Externally set output



bility may be improved by increasing C2 to 2.2µF and reducing R2 to  $10k\Omega$ . See "Component Selection -General" for input capacitor requirements.

#### **Thermal Considerations**

The worst-case power dissipation for this part is given

$$P_{D(MAX)} = (VIN_{(MAX)} - VO_{(MIN)}) \bullet I_{O(MAX)} + VIN_{(MAX)} \bullet I_{Q(MAX)}$$

For all practical purposes, it can be reduced to:

$$P_{D(MAX)} = (VIN_{(MAX)} - VO_{(MIN)}) \cdot I_{O(MAX)}$$

Looking at a typical application:

 $VIN_{(MAX)} = 4.2V$ 

VO = 3V - 3.5% worst-case

 $I_O = 150$ mA  $T_A = 85$ °C

$$P_{D(MAX)} = (4.2 - 2.895) \cdot 0.150 = 196 \text{ mW}$$

Using this figure, we can calculate the maximum thermal impedance allowable to maintain  $T_J \le 150$ °C:

$$\theta_{(J-A)(MAX)} = \frac{\left(T_{J(MAX)} - T_{A(MAX)}\right)}{P_{D(MAX)}} = \frac{\left(150 - 85\right)}{0.196} = 332 \,^{\circ}\text{C/W}$$

With the standard SOT-23-5 Land Pattern shown at the end of this datasheet, and minimum trace widths, the thermal impedance junction to ambient for SC8863 is 256°C/W. Thus with no additional heatsinking,  $T_{J(MAX)} = 135^{\circ}C.$ 

The junction temperature can be further reduced by the use of larger trace widths, and connecting pcb copper area to the GND pin (pin 2), which connectes directly to the device substrate. Adding approximately one square inch of pcb copper to pin 2 will reduce  $\theta_{TH(J-A)}$  to approximately 130°C/W and  $T_{J(MAX)}$  to approximately 110°C, for example. Lower junction temperatures improve overall output voltage accuracy. A sample pcb layout for the Internally Preset Output Voltage circuit on page 2 is shown in Figure 2 below.

#### **Layout Considerations**

While layout for linear devices is generally not as critical as for a switching application, careful attention to detail will ensure reliable operation. See Figure 2 below for a sample layout.

- 1) Attaching the part to a larger copper footprint will enable better heat transfer from the device, especially on PCBs where there are internal ground and power planes.
- 2) Place the input and output capacitors (and the capacitor from OUT to SET for adjustable applications) close to the device for optimal transient response and device behaviour.
- 3) Connect all ground connections directly to the ground plane. If there is no ground plane, connect to a common local ground point before connecting to board ground.

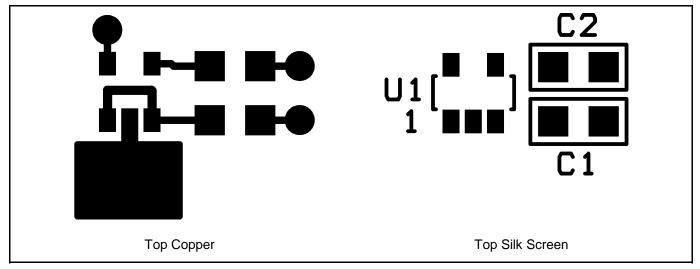
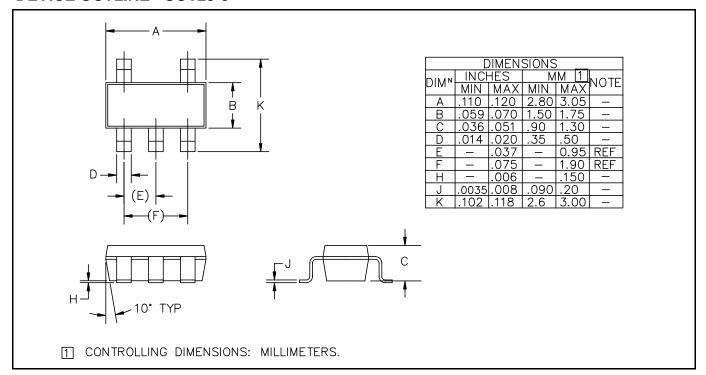


Figure 2: Suggested pcb layout based upon internally preset output voltage application on page 2. NOTES:

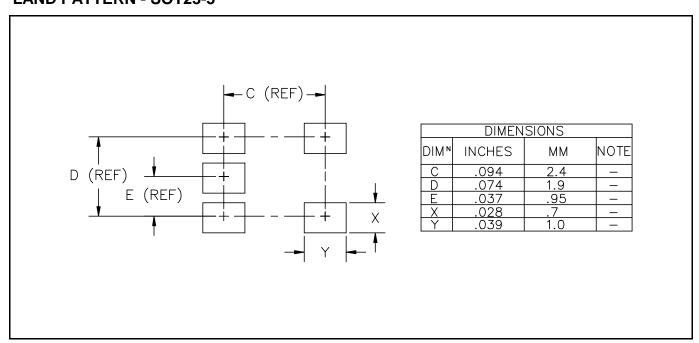
- (1) All vias go to the ground plane.
- (2) Copper area on pin 2 is recommended, but not required. Connect to the ground plane with a via or vias.



### **DEVICE OUTLINE - SOT23-5**



#### **LAND PATTERN - SOT23-5**



ECN 00-915